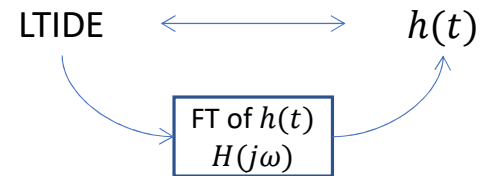


# Signals and Systems I

Topic 11

Laplace Transform

# Fourier Transform and LTIDE



Reminder:

What is the output of the following system to  $x(t)$ ?

$$\frac{d^2 y(t)}{dt^2} + 5 \frac{dy(t)}{dt} + 4y(t) = \frac{dx(t)}{dt} - 6x(t)$$

First you need to find  $h(t)$  that is the “impulse response” of this system and the output is  $y(t) = x(t) * h(t)$ .

Alternative approach using the FT is to first find  $H(j\omega)$  from the equation and then calculate inverse FT of  $H(j\omega)X(j\omega)$  which is  $y(t)$ .

This method can be used for cases where  $M$  (order of highest derivative of input) is even greater than  $N$  (system order that is the order of highest derivative of output):

What is  $h(t)$  impulse response of this system?

$$\frac{dy(t)}{dt} + 4y(t) = \frac{d^2 x(t)}{dt} + \frac{dx(t)}{dt} - 6x(t)$$

# Laplace Transform

Fourier Transform provides a linear combination of  $e^{j\omega t}$ s for a signal:

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega$$

Laplace Transform is an extension of this method and finds correlation of  $e^{st}$  where  $s = \sigma + j\omega$  (in FT  $s = j\omega$  only), with the signal.

$$X(j\omega) = \underbrace{\int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt}_{s=j\omega} \xrightarrow{\text{extension to Laplace}} X(s) = \underbrace{\int_{-\infty}^{\infty} x(t) e^{-st} dt}_{s=j\omega+\sigma}$$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega \xrightarrow{\text{extension to Laplace}} x(t) = \frac{1}{2\pi j} \underbrace{\int_{c-\infty}^{c+\infty} X(s) e^{st} ds}_{s=c+j\omega, c \text{ is a constant}} \quad \text{set } c \text{ to zero and its FT}$$

Similar to FT, Laplace is a linear operation.

$$\begin{aligned} x_1(t) &\rightarrow X_1(s) \\ x_2(t) &\rightarrow X_2(s) \\ ax_1(t) + bx_2(t) &\rightarrow aX_1(s) + bX_2(s) \end{aligned}$$

Why Laplace?

Many signals don't have FT  
but have Laplace Transform!

# Laplace Transform

Fourier Transform provides a linear combination of  $e^{j\omega t}$ s for a signal:

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega$$

Laplace Transform is an extension of this method and finds correlation of  $e^{st}$  where  $s = \sigma + j\omega$  (in FT  $s = j\omega$  only), with the signal!

$$X(j\omega) = \underbrace{\int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt}_{s=j\omega} \xrightarrow{\text{extension to Laplace}} X(s) = \underbrace{\int_{-\infty}^{\infty} x(t) e^{-st} dt}_{s=j\omega+\sigma}$$



# Region of Convergence (ROC) of Laplace Transform

Set of  $s$  values in complex plane that makes the integral of  $X(s)$  converge.

Example:  $x(t) = e^{-at}u(t)$ ,  $a > 0$ . Find  $X(s)$  & its ROC.

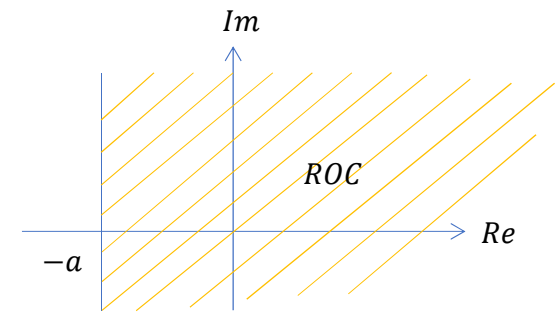
$$\begin{aligned} X(s) &= \int_{-\infty}^{\infty} e^{-at}u(t)e^{-st}dt = \int_0^{\infty} e^{-(a+s)t}dt \\ &= \frac{e^{-(a+s)t}}{(a+s)} \Big|_0^{\infty} = \frac{e^{-(a+s)\infty}}{-(a+s)} + \frac{1}{a+s} \end{aligned}$$

$$\begin{aligned} \lim_{t \rightarrow \infty} \frac{e^{-(a+s)t}}{-(a+s)} &= \lim_{t \rightarrow \infty} \frac{e^{-(a+\sigma+j\omega)t}}{-(a+s)} \\ &= \lim_{t \rightarrow \infty} \frac{e^{-j\omega t} e^{-(a+\sigma)t}}{-(a+s)} \end{aligned}$$

$e^{-j\omega t}$  is a complex number,  $|e^{-j\omega t}| = 1$  for all  $t$ .

$$= \lim_{t \rightarrow \infty} \frac{e^{-(a+\sigma)t}}{-(a+s)} = \begin{cases} 0 & a + \sigma > 0 \text{ converges (ROC)} \\ \infty & a + \sigma < 0 \text{ does not converge} \end{cases}$$

$$\Rightarrow \text{ROC} = \{“s = \sigma + j\omega” \text{ such that } \sigma + a > 0 \Rightarrow \sigma > -a\}$$



$$x(t) = e^{-at}u(t), \quad a > 0 \quad \xrightarrow{L} \quad \frac{1}{s+a}$$

$$\text{ROC} = \{s, \text{Re}\{s\} > -a, \quad a > 0\}$$

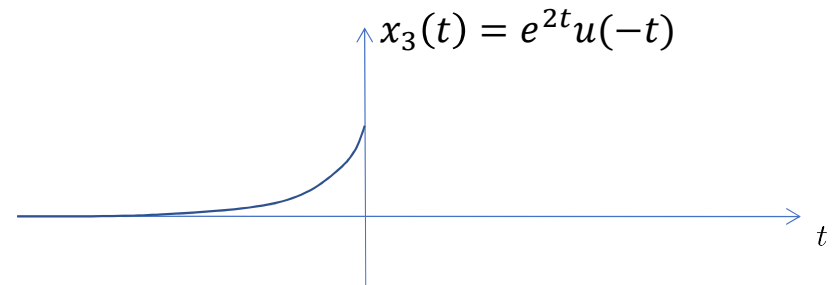
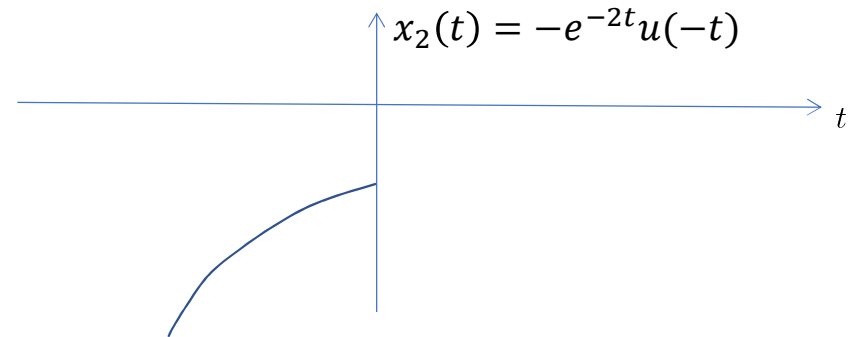
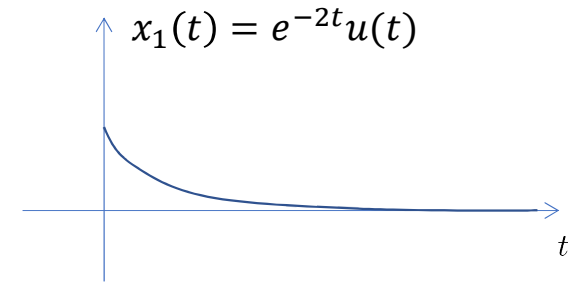
If  $s = j\omega$  is in ROC then FT exists

# Laplace Transform & ROC

## Example:

Find Laplace and ROC for the following signals:

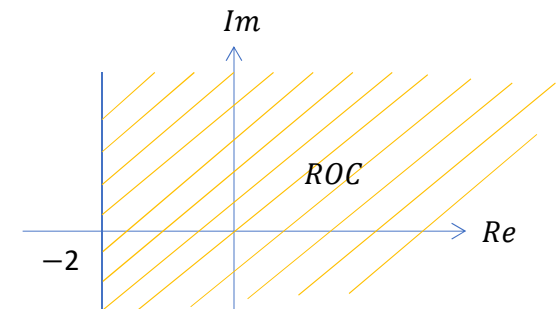
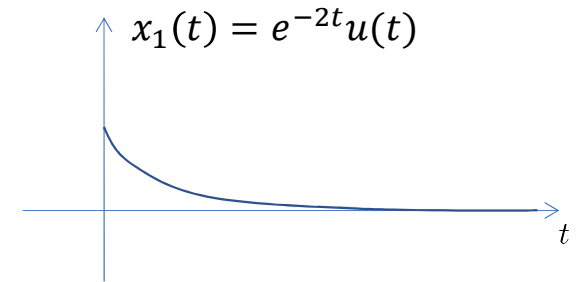
$$X(s) = \int_{-\infty}^{\infty} x(t)e^{-st} dt$$



## Laplace Transform & ROC

$$X_1(s) = \int_0^{\infty} e^{-(2+s)t} dt = \frac{1}{2+s} + \underbrace{\frac{e^{-(2+s)\infty}}{-(2+s)}}_{\substack{0 \text{ iff } 2+\sigma > 0}}$$

$$\text{ROC} = \{s \mid \text{Re}\{s\} > -2\} \rightarrow X_1(s) = \frac{1}{2+s}$$



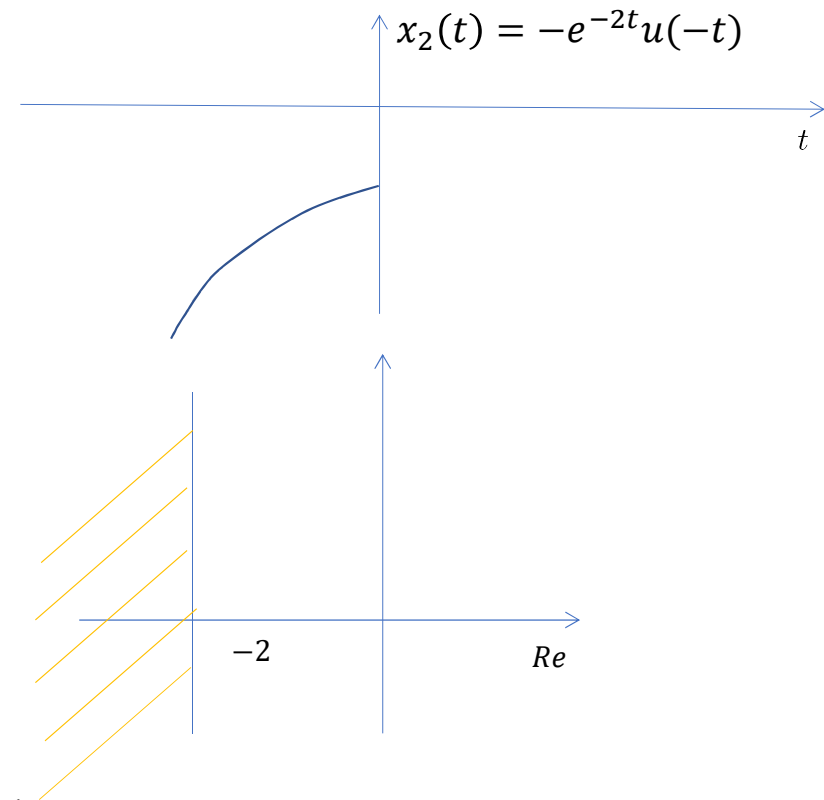
\*Note: Direction of the dashed area of ROC is the same as direction of the signal in time.

## Laplace Transform & ROC

$$\begin{aligned}
 X_2(s) &= \int_{-\infty}^0 -e^{-2t} e^{-st} dt = - \int_{-\infty}^0 e^{-(2+s)t} dt \\
 &= \frac{-e^{-(2+s)t}}{-(2+s)} \Big|_{-\infty}^0 = \frac{1}{(2+s)} + \frac{e^{-(2+s)(-\infty)}}{-(2+s)}
 \end{aligned}$$

$$e^{(2+s)(\infty)} = \begin{cases} 0 & 2 + \sigma < 0 \rightarrow \text{ROC} = \text{Re}\{s\} < -2 \\ \infty & 2 + \sigma > 0 \end{cases}$$

$$X_2(s) = \frac{1}{2+s}, \quad \text{Re}\{s\} < -2$$



$s = j\omega$  is not in ROC. Signal doesn't have FT.

Laplace is the same as that of  $x_1(t)$ , only ROC is different!

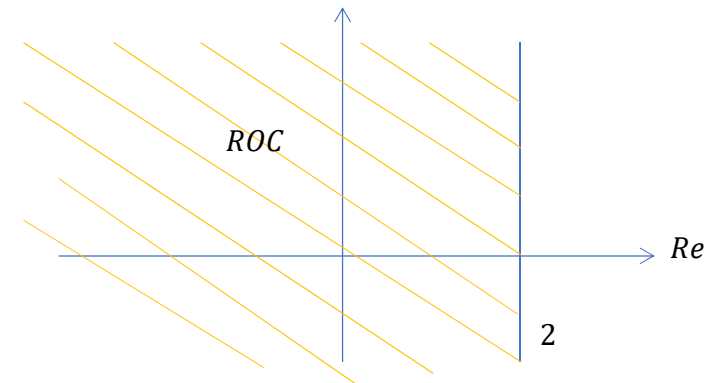
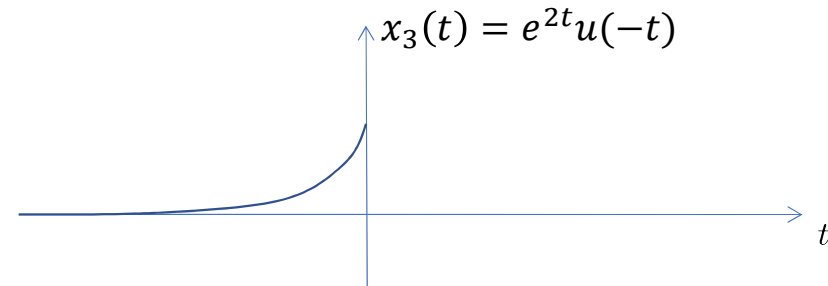


## Laplace Transform & ROC

$$\begin{aligned} X_3(s) &= \int_{-\infty}^0 e^{2t} e^{-st} dt = \int_{-\infty}^0 e^{(2-s)t} dt \\ &= \frac{e^{(2-s)t}}{(2-s)} \Big|_{-\infty}^0 = \frac{1}{(2-s)} - \frac{e^{-(2-s)(\infty)}}{(2-s)} \end{aligned}$$

$$e^{-(2-s)(\infty)} = \begin{cases} 0 & 2 - \sigma > 0 \rightarrow \text{ROC} = \text{Re}\{s\} < 2 \\ \infty & 2 - \sigma < 0 \end{cases}$$

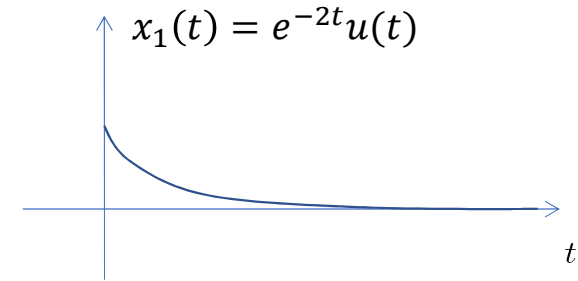
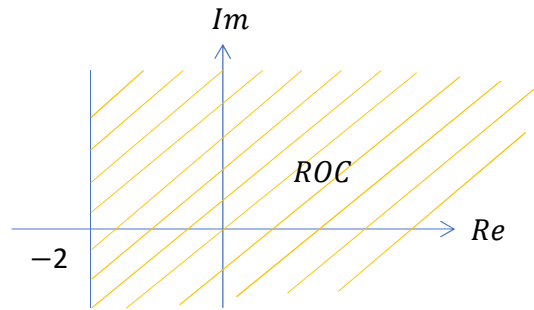
$$X_3(s) = \frac{1}{2-s}, \quad \text{Re}\{s\} < 2$$



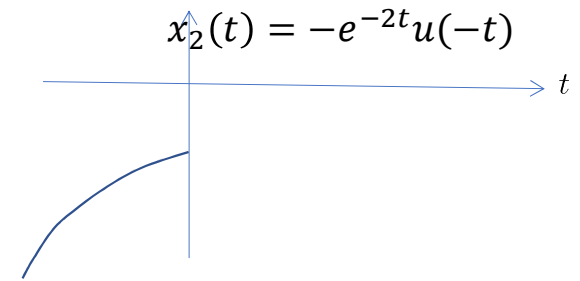
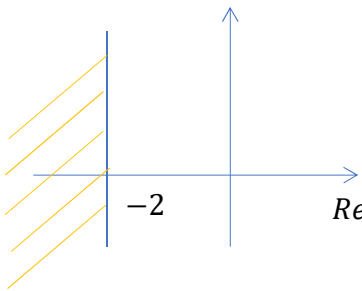
$s = j\omega$  is in ROC. This noncausal signal has FT.

# Laplace Transform & ROC

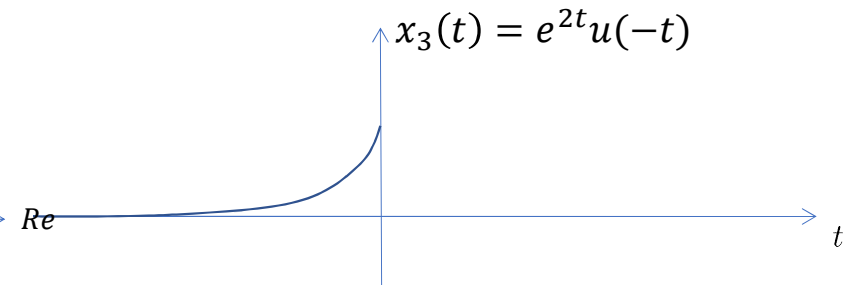
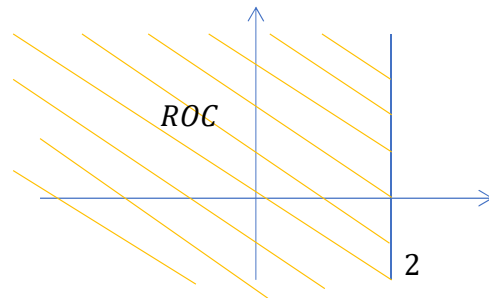
$$X_2(s) = \frac{1}{2+s}, \quad \text{Re}\{s\} > -2$$



$$X_2(s) = \frac{1}{2+s}, \quad \text{Re}\{s\} < -2$$



$$X_3(s) = \frac{1}{2-s}, \quad \text{Re}\{s\} < 2$$



## Laplace Transform & ROC

### Example:

Find  $X(s)$  for  $e^{-a|t|}$ ,  $a > 0$ .

$$x(t) = e^{-a|t|} = e^{-at}u(t) + e^{at}u(-t)$$

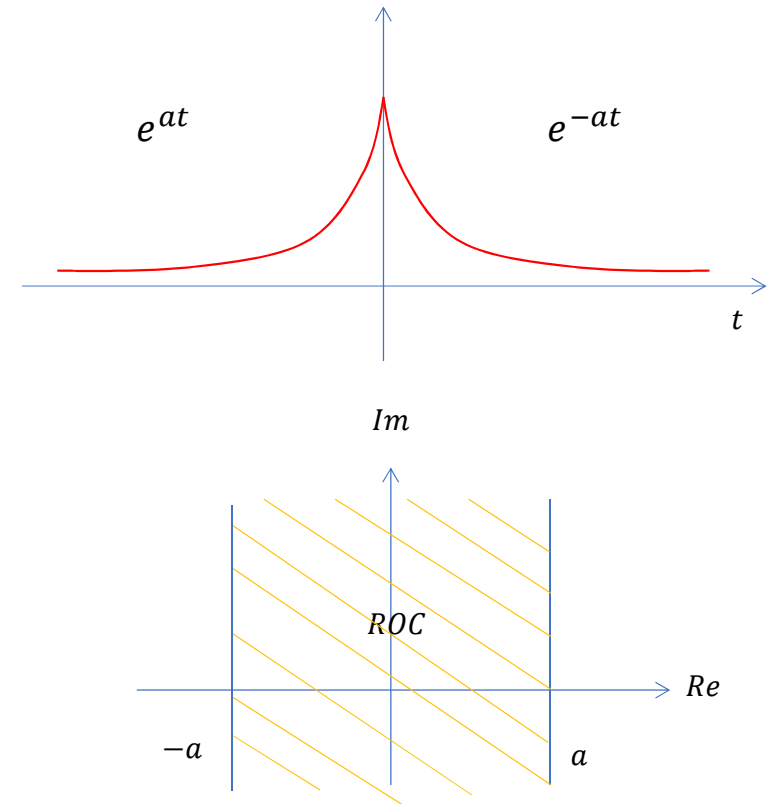
$$X(s) = \mathcal{L} (e^{-at}u(t)) + \mathcal{L} (e^{at}u(-t))$$

$$ROC (X(s)) = ROC\{\mathcal{L} (e^{-at}u(t))\} \cap ROC\{\mathcal{L} (e^{at}u(-t))\}$$

$$e^{-at}u(t) \xrightarrow{\mathcal{L}} \frac{1}{a+s}, \operatorname{Re}\{s\} > -a$$

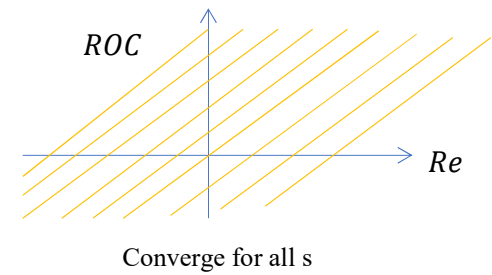
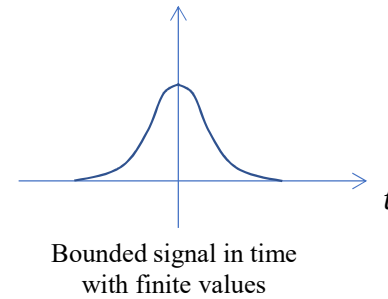
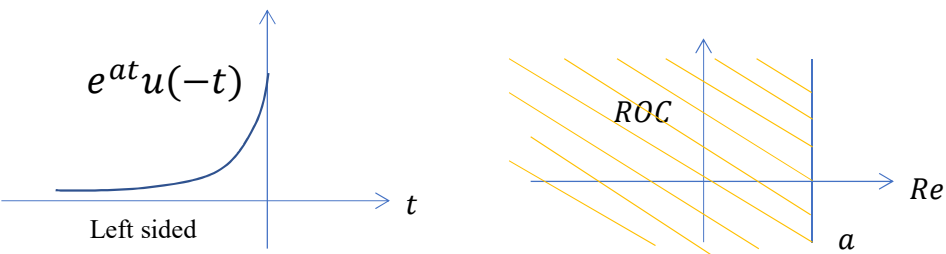
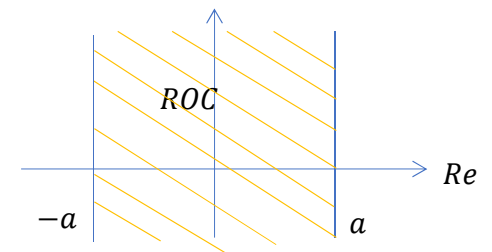
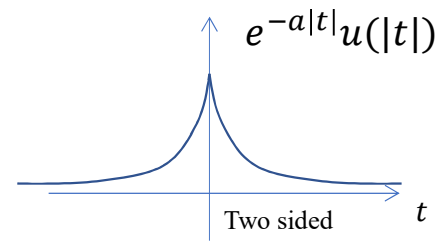
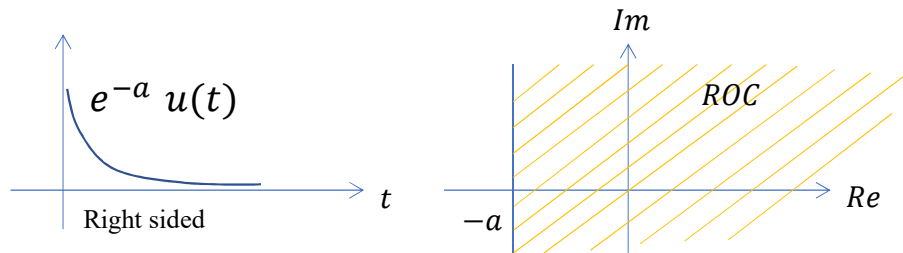
$$e^{at}u(-t) \xrightarrow{\mathcal{L}} \frac{1}{a-s}, \operatorname{Re}\{s\} < a$$

$$X(s) = \frac{1}{a+s} + \frac{1}{a-s} = \frac{2a}{a^2 - s^2}$$



# Laplace Transform & ROC

Left sided, Right sided or double sided signals and their ROC



## Laplace Transform & ROC

### **Example:**

Find  $x(t)$  for the following  $X(s)$ .

$$X(s) = \frac{1}{s+2} + \frac{1}{s+3}, \quad \text{ROC} = \{s \mid \text{Re}(s) > -2\}$$

$$x(t) = e^{-2t}u(t) + e^{-3t}u(t)$$

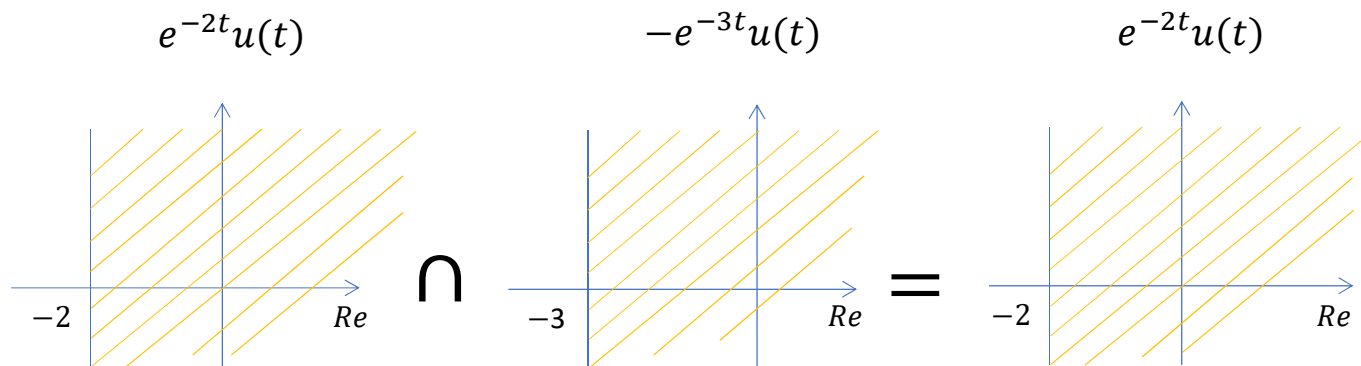
## Laplace Transform & ROC

### Example:

Find  $x(t)$  for the following  $X(s)$ .

$$X(s) = \frac{1}{s+2} + \frac{1}{s+3}, \quad \text{ROC} = \{s \mid \text{Re}(s) > -2\}$$

$$x(t) = e^{-2t}u(t) + e^{-3t}u(t)$$



## Laplace Transform & ROC

### **Example:**

Find  $x(t)$  for the following  $X(s)$ .

$$X(s) = \frac{1}{s+2} + \frac{1}{s+3}, \quad ROC = \{s \mid -3 < Re(s) < -2\}$$

## Laplace Transform & ROC

### Example:

Find  $x(t)$  for the following  $X(s)$ .

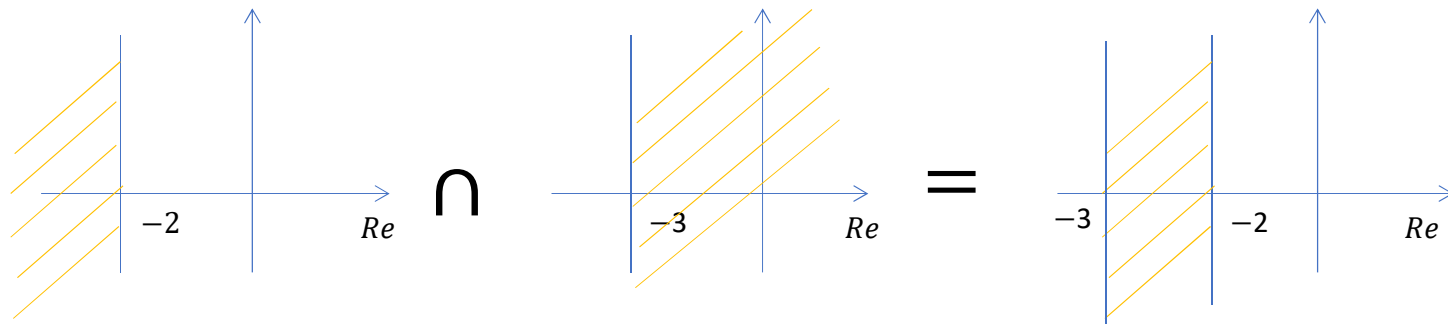
$$X(s) = \frac{1}{s+2} + \frac{1}{s+3}, \quad \text{ROC} = \{s \mid -3 < \text{Re}(s) < -2\}$$

Signal has to be two-sided

$$x(t) = -e^{-2t}u(-t) + e^{-3t}u(t)$$

$$-e^{-2t}u(-t)$$

$$e^{-3t}u(t)$$



Without ROC there are more than one option for the inverse of the Laplace transform.



# Laplace Transform & ROC

## Example:

Find possible  $x(t)$ s with the following Laplace transform:

$$X(s) = \frac{1}{(s+2)(s+3)}$$

Use P.F.E

$$X(s) = \frac{a}{s+2} + \frac{b}{s+3} = \frac{1}{(s+2)(s+3)}$$

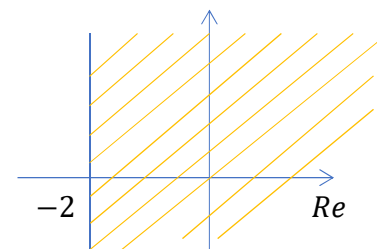
$$a(s+3) + b(s+2) = 1 \rightarrow \begin{cases} \text{set } s = -3 \Rightarrow b = -1 \\ \text{set } s = -2 \Rightarrow a = 1 \end{cases}$$

$$X(s) = \frac{1}{s+2} + \frac{-1}{s+3}$$

Possible ROCs for

$$\frac{1}{s+2}$$

$$e^{-2t}u(t)$$

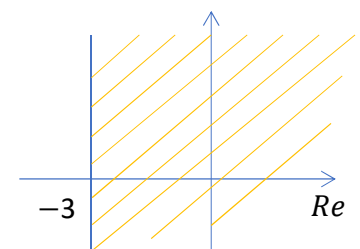


(a)

Possible ROCs for

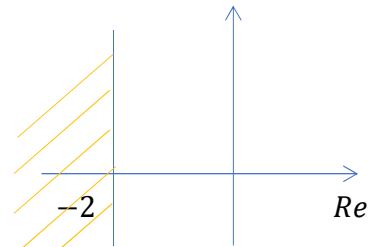
$$\frac{-1}{s+3}$$

$$-e^{-3t}u(t)$$



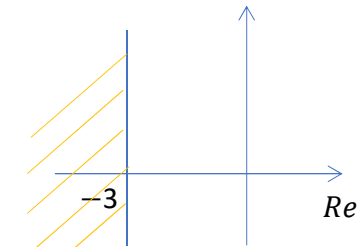
(b)

$$-e^{-2t}u(-t)$$



(c)

$$e^{-3t}u(-t)$$

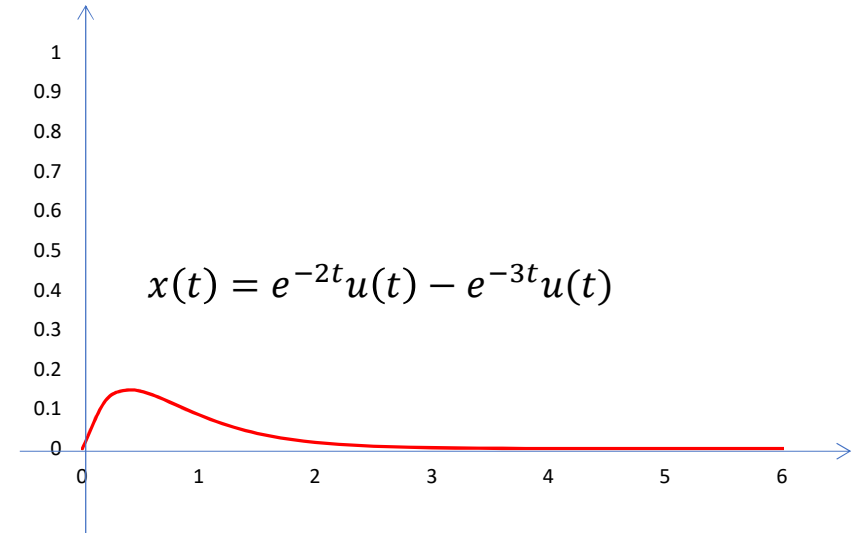
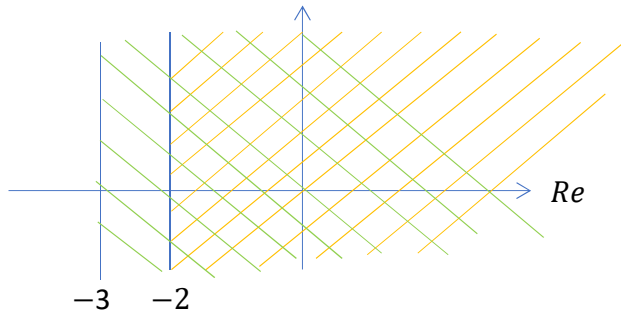


(d)

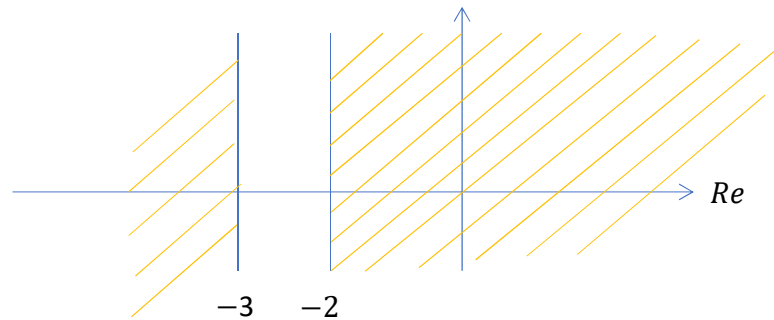
# Laplace Transform & ROC

Case 1: (a) & (b):

Intersection:  $Re\{s\} > -2$



Case 2: (a) & (d):

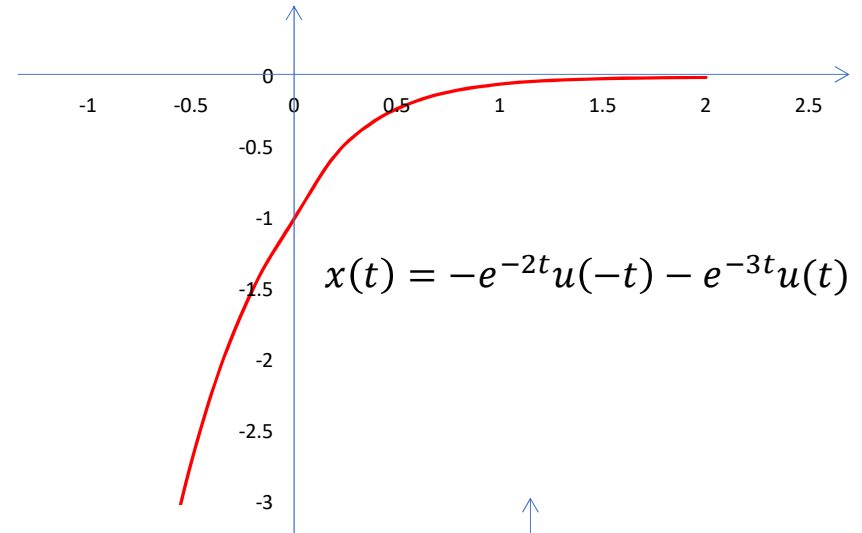
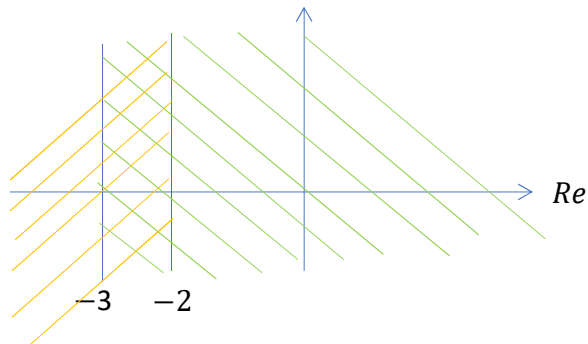


No intersection, so this is a signal that has no Laplace transform!

# Laplace Transform & ROC

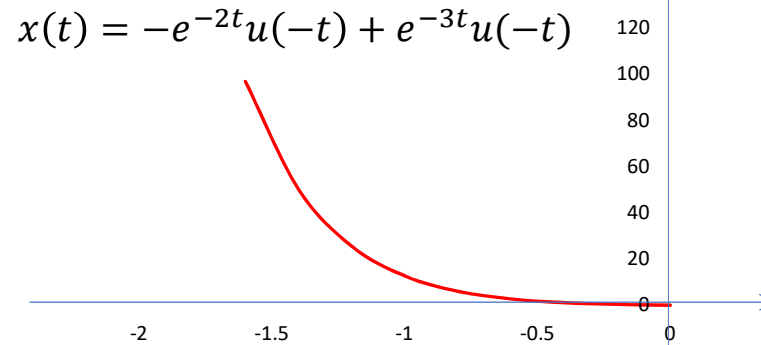
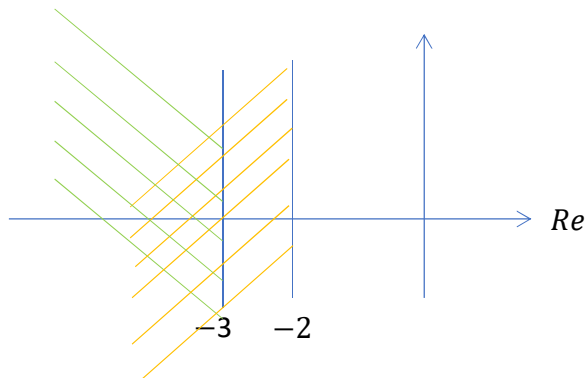
Case 3: (b) & (c):

Intersection:  $-3 < \text{Re}\{s\} < -2$



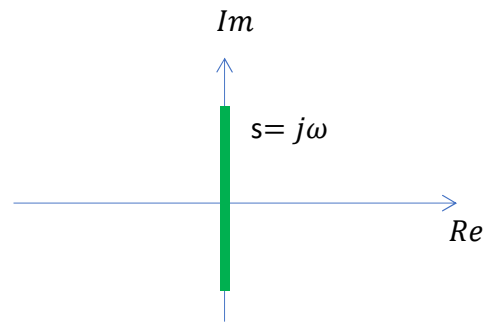
Case 4: (c) & (d):

Intersection:  $\text{Re}\{s\} < -3$



## Laplace Transform & Fourier Transform

FT is a Laplace transform that is calculated for  $s = j\omega$ .



Therefore a signal has FT only if ROC includes the  $j\omega$  axis! Therefore, **Not all signals have FT.**

Question: Go back to the previous example and indicate which signal has FT?

## Laplace Transform & ROC

### **Example:**

Find  $x(t)$  with the following  $X(s)$ :

$$X(s) = \frac{2s^2+5}{s^2+3s+2} \text{ and } ROC : Re\{s\} > -1 \text{ (Right sided signal)}$$

## Laplace Transform & ROC

### Example:

Find  $x(t)$  with the following  $X(s)$ :

$$X(s) = \frac{2s^2+5}{s^2+3s+2} \text{ and } ROC : Re\{s\} > -1 \text{ (Right sided signal)}$$

Solution:

$$X(s) = \frac{2(s^2 + 3s + 2) + 1 - 6s}{s^2 + 3s + 2} = 2 + \frac{1 - 6s}{s^2 + 3s + 2}$$

$$P.F.E \left( \frac{1 - 6s}{s^2 + 3s + 2} \right) = \frac{a}{(s + 1)} + \frac{b}{(s + 2)} = \frac{1 - 6s}{(s + 1)(s + 2)}$$

$$a = \left. \frac{1 - 6s}{s + 2} \right|_{s=-1} = 7$$

$$b = \left. \frac{1 - 6s}{s + 1} \right|_{s=-2} = -13$$

$$x(t) = 2\delta(t) + 7e^{-t}u(t) - 13e^{-2t}u(t)$$

# Laplace Transform Properties

Linearity  $ax_1(t) + bx_2(t) \rightarrow aX_1(s) + bX_2(s)$

Time Shift  $x(t - t_0) \rightarrow e^{-st_0} X(s)$

Frequency shift  $x(t)e^{s_0 t} \rightarrow X(s - s_0)$

Derivative  $\frac{dx(t)}{dt} \rightarrow sX(s)$

Higher Order Derivative  $\frac{d^n x(t)}{dt^n} \rightarrow s^n X(s)$

$$\int_{-\infty}^t x(t) dt \rightarrow \frac{1}{s} X(s)$$

Scaling  $x(at) \rightarrow \frac{1}{a} X\left(\frac{s}{a}\right)$

$$x_1(t) * x_2(t) \rightarrow X_1(s) X_2(s)$$

$$x_1(t) \times x_2(t) \rightarrow \frac{1}{2\pi} X_1(s) * X_2(s)$$

Bilateral Laplace Transform:

$$X(s) = \int_{-\infty}^{\infty} x(t) e^{-st} dt$$

# Laplace Transform Properties

Linearity  $ax_1(t) + bx_2(t) \rightarrow aX_1(s) + bX_2(s)$

Time Shift  $x(t - t_0) \rightarrow e^{-st_0} X(s)$

Frequency shift  $x(t)e^{s_0 t} \rightarrow X(s - s_0)$

Derivative  $\frac{dx(t)}{dt} \rightarrow sX(s) - x(0)$

Higher Order Derivative  $\frac{d^n x(t)}{dt^n} \rightarrow s^n X(s) - \sum_{k=1}^n s^{n-k} x^{(k-1)}(0)$

$$\int_{-\infty}^t x(t) dt \rightarrow \frac{1}{s} X(s)$$

Scaling  $x(at) \rightarrow \frac{1}{a} X\left(\frac{s}{a}\right) \quad a > 0$

$$x_1(t) * x_2(t) \rightarrow X_1(s) X_2(s)$$

$$x_1(t) \times x_2(t) \rightarrow \frac{1}{2\pi} X_1(s) * X_2(s)$$

Initial value Theorem  $x(0) = \lim_{s \rightarrow \infty} sX(s)$

Final value Theorem  $x(\infty) = \lim_{s \rightarrow 0} sX(s)$

Bilateral Laplace Transform:

$$X(s) = \int_{-\infty}^{\infty} x(t)e^{-st} dt$$

Unilateral Laplace Transform:

$$X(s) = \int_0^{\infty} x(t)e^{-st} dt$$

Unilateral Laplace is used for Causal signals and causal systems to deal with **initial conditions**.



# Laplace Transform Properties

**Example:**

$$\begin{aligned}h(t) &= \delta(t - 3) \\H(s) &= \int_{-\infty}^{\infty} \delta(t - 3)e^{-ts} dt \\&= e^{-3s} \underbrace{\int_{-\infty}^{\infty} \delta(t - 3) dt}_1 \\&= e^{-3s}\end{aligned}$$

**Example:**

$$\begin{aligned}X(s) &= \frac{e^{-3s}}{s} = \frac{1}{s} \times e^{-3s} \\x(t) &= \underbrace{u(t)}_{\frac{1}{s}} * \underbrace{\delta(t - 3)}_{e^{-3s}} = u(t - 3)\end{aligned}$$

Delay of  $u(t)$  by 3 or  
Integral of  $\delta(t - 3)$

Useful Laplace Transforms:

$$\begin{aligned}e^{-at} \cos(bt)u(t) &\xrightarrow{\text{Laplace}} \frac{s+a}{(s+a)^2+b^2} \\ \delta(t) &\xrightarrow{\text{Laplace}} 1 \\ u(t) &\xrightarrow{\text{Laplace}} \frac{1}{s} \\ tu(t) &\xrightarrow{\text{Laplace}} \frac{1}{s^2}\end{aligned}$$

**Example:**

$$X(s) = \frac{1}{s(s+5)} = \frac{a}{s} + \frac{b}{s+5}, \quad ROC = Re\{s\} > 0$$

$$a = \left. \frac{1}{s+5} \right|_{s=0} = \frac{1}{5}$$

$$b = \left. \frac{1}{s} \right|_{s=-5} = \frac{-1}{5}$$

$$x(t) = \frac{1}{5}u(t) - \frac{1}{5}e^{-5t}u(t)$$

# Laplace Transform Properties

Example:

$$\begin{aligned}x(t) &= e^{-2t} \cos\left(\frac{\pi}{3}t\right)u(t) \\&= e^{-2t} \left( \frac{e^{j\frac{\pi}{3}t} + e^{-j\frac{\pi}{3}t}}{2} \right) u(t) \\&= \underbrace{\frac{e^{j\frac{\pi}{3}t}}{2}}_{\frac{1}{2}e^{s_0 t}} \underbrace{e^{-2t}u(t)}_{\frac{1}{s+2}} + \frac{e^{-j\frac{\pi}{3}t}}{2} e^{-2t}u(t) \\&= \frac{1}{2} \frac{1}{\underbrace{\left(s - j\frac{\pi}{3}\right)}_{s_0} + 2} + \frac{1}{2} \frac{1}{\left(s + j\frac{\pi}{3}\right) + 2} \\&= \frac{1}{2} \left[ \frac{1}{\left(s - j\frac{\pi}{3}\right) + 2} + \frac{1}{\left(s + j\frac{\pi}{3}\right) + 2} \right] \\&= \frac{1}{2} \left[ \frac{s + j\frac{\pi}{2} + 2 + s - j\frac{\pi}{2} + 2}{\left[\left(s - j\frac{\pi}{3}\right) + 2\right] \left[\left(s + j\frac{\pi}{3}\right) + 2\right]} \right] \\&= \frac{1}{2} \left[ \frac{2s + 4}{\left[\left(s + 2\right) - j\frac{\pi}{3}\right] \left[\left(s + 2\right) + j\frac{\pi}{3}\right]} \right] \\&= \frac{s + 2}{\left(s + 2\right)^2 + \left(\frac{\pi}{3}\right)^2} = \frac{s + 2}{s^2 + 4s + \left(4 + \frac{\pi^2}{9}\right)}\end{aligned}$$

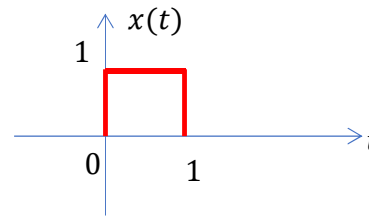
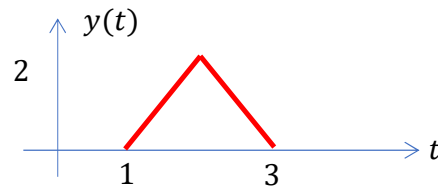
Alternatively use:

$$e^{s_0 t}u(t), s_0 = \alpha_0 + j\omega_0 \rightarrow X_1(s) = \frac{1}{s - s_0}, ROC = Re\{s\} > \alpha_0$$

## Laplace Transform Properties

**Example:**

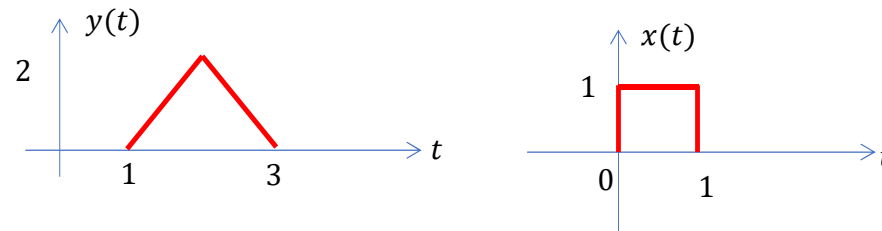
Write  $Y(s)$  as a function of  $X(s)$ .



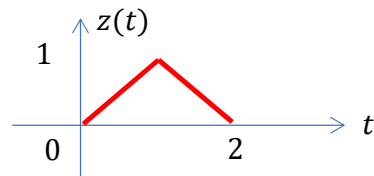
## Laplace Transform Properties

**Example:**

Write  $Y(s)$  as a function of  $X(s)$ .



Solution:

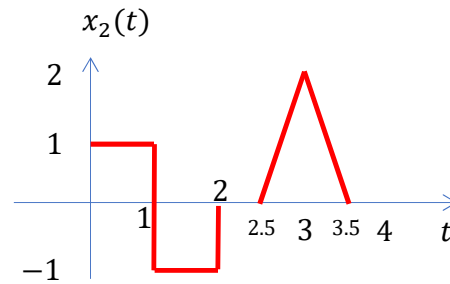
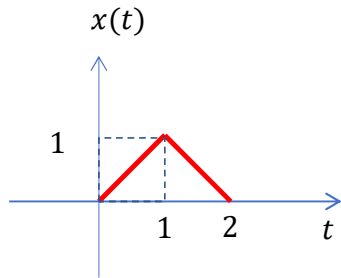


$$x(t) * x(t) = z(t) \Rightarrow X(s) \times X(s) = Z(s)$$

$$2z(t - 1) = y(t) \Rightarrow Y(s) = 2e^{-s}Z(s) = 2e^{-s}X^2(s)$$

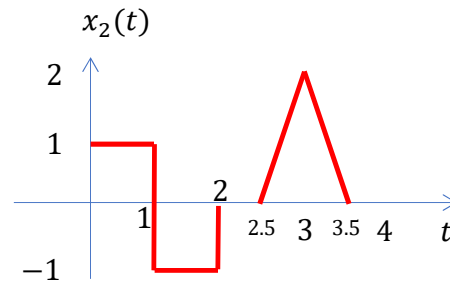
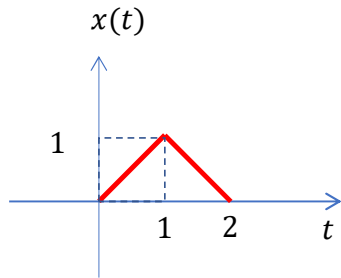
## Laplace Transform Properties

**Example:** Write Laplace transform of  $x_2(t)$  as a function of  $X(s)$ .

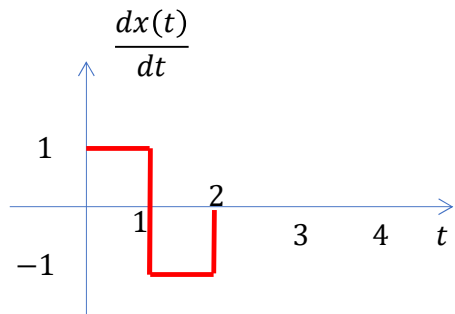


# Laplace Transform Properties

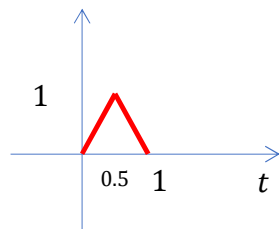
**Example:** Write Laplace transform of  $x_2(t)$  as a function of  $X(s)$ .



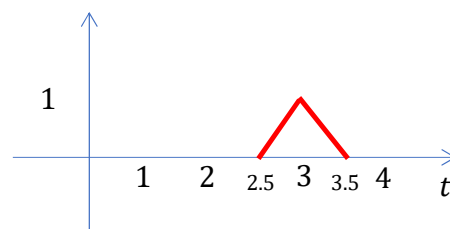
$\geq$



$$z(t) = x(2t)$$



$$z(t - 2.5) = x(2(t - 2.5)) = x(2t - 5)$$



$$\begin{aligned} x_2(t) &= \frac{dx(t)}{dt} + 2x(2t - 5) \\ X_2(s) &= sX(s) + 2\mathcal{L}\{x(2t - 5)\} \\ &= sX(s) + 2\mathcal{L}\{z(t - 2.5)\} \\ &= sX(s) + 2e^{-2.5s} \mathcal{L}\{z(t)\} \\ &= sX(s) + 2e^{-2.5s} \mathcal{L}\{x(2t)\} \\ &= sX(s) + 2e^{-2.5s} \frac{1}{2} X\left(\frac{s}{2}\right) \end{aligned}$$

# Laplace Transform

## **Example:**

The input and output of a casual LTI system respectively are:  $x(t) = e^{-2t}u(t)$  and  $y(t) = te^{-t}u(t)$ . Find  $H(s)$ , Laplace transform of the impulse response  $h(t)$  and show its ROC.

# Laplace Transform

## Example:

The input and output of a casual LTI system respectively are:  $x(t) = e^{-2t}u(t)$  and  $y(t) = te^{-t}u(t)$ . Find  $H(s)$ , Laplace transform of the impulse response  $h(t)$  and show its ROC.

## Solution:

$$y(t) = x(t) * h(t) \rightarrow Y(s) = X(s)H(s) \rightarrow H(s) = \frac{Y(s)}{X(s)}$$

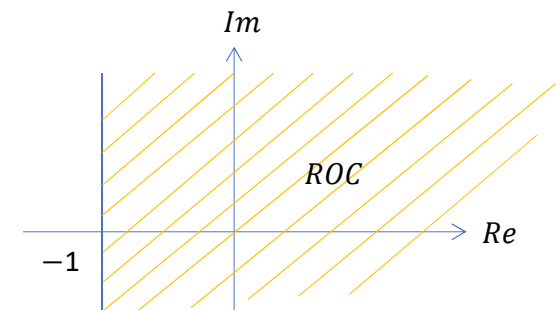
$$x(t) = e^{-2t}u(t) \rightarrow X(s) = \frac{1}{s+2}$$

$$y(t) = te^{-t}u(t) \rightarrow Y(s) = \frac{1}{(s+1)^2}$$

$$H(s) = \frac{Y(s)}{X(s)} = \frac{\frac{1}{(s+1)^2}}{\frac{1}{(s+2)}} = \frac{s+2}{(s+1)^2}$$

Since the LTI system is casual,  $h(t)$  is right hand signal, therefore ROC:  $Re\{s\} > -1$

Laplace of impulse response of an LTI system is also known as **Transfer Function** of the system.





# Laplace Transform

## Example:

LTI system has transform  $H(s) = \frac{s-1}{s+1}$ , what is  $y(t)$  output of this system to input with Laplace transform  $X(s) = \frac{s}{s+1}$ .

# Laplace Transform

## Example:

LTI system has transform  $H(s) = \frac{s-1}{s+1}$ , what is  $y(t)$  output of this system to input with Laplace transform  $X(s) = \frac{s}{s+1}$ .

## Solution:

$$\begin{aligned} Y(s) &= H(s) \times X(s) = \frac{s-1}{s+1} \times \frac{s}{s+1} = \frac{s(s-1)}{(s+1)^2} \\ &= \frac{s^2 - s}{s^2 + 2s + 1} = 1 + \frac{-3s - 1}{s^2 + 2s + 1} = 1 + \frac{a}{s+1} + \frac{b}{(s+1)^2} \end{aligned}$$

$$b = (s+1)^2 Y(s) \Big|_{s=-1} = 2$$

$$a = \frac{d}{ds} ((s+1)^2 Y(s)) \Big|_{s=-1} = (2s-1) \Big|_{s=-1} = -3$$

$$\frac{1}{(s+1)^2} \xrightarrow{IL} te^{-t}u(t)$$

$$y(t) = \delta(t) - 3e^{-t}u(t) + 2te^{-t}u(t)$$

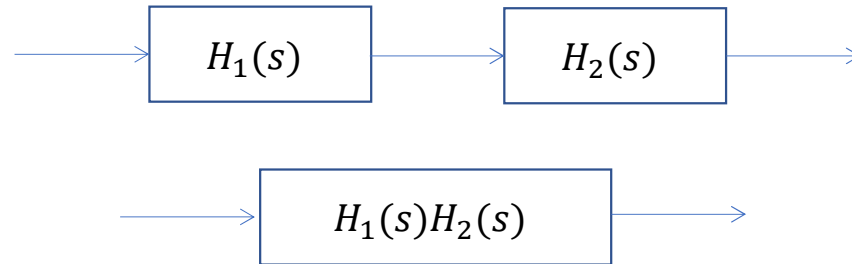
Another method:

$$Y(s) = \frac{s^2 - s}{(s+1)^2} = \underbrace{\frac{s^2}{(s+1)^2}}_{\left(\frac{d^2}{dt^2}(te^{-t}u(t))\right)} - \underbrace{\frac{s}{(s+1)^2}}_{\left(\frac{d}{dt}(te^{-t}u(t))\right)}$$

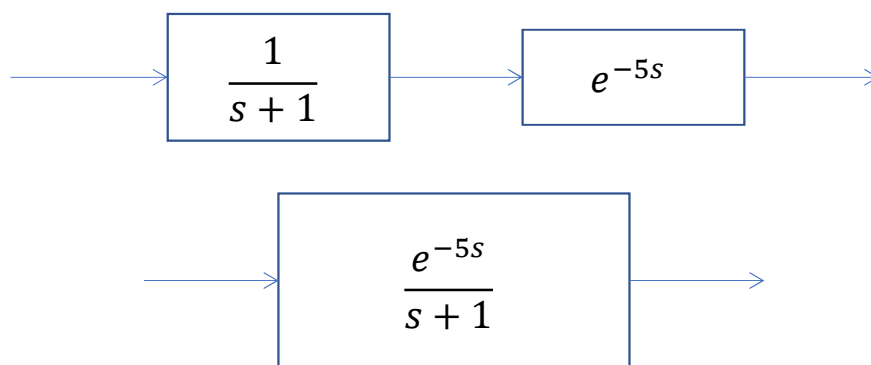
$$\begin{aligned} y(t) &= \frac{d}{dt} [(e^{-t} - te^{-t})u(t)] - te^{-t}\delta(t) + (e^{-t} - te^{-t})u(t) \\ &= \delta(t) + (2te^{-t} - 3e^{-t})u(t) \end{aligned}$$

# Laplace Transform

## Cascade



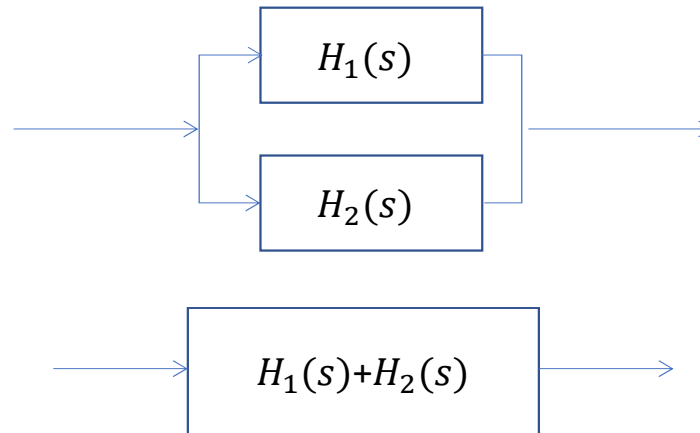
## Example:



$$H(s) = \frac{e^{-5s}}{s+1} \rightarrow h(t) = e^{-(t-5)}u(t-5)$$
$$Z(s) = \frac{1}{s+1} \rightarrow z(t) = e^{-t}u(t)$$

# Laplace Transform

## Parallel



Example:

$$H_1(s) = \frac{1}{s + 2 - 4j}, \quad H_2(s) = \frac{1}{s + 2 + 4j}$$

$$H(s) = \frac{1}{s + 2 - 4j} + \frac{1}{s + 2 + 4j} = \frac{s + 2 + 4j + s + 2 - 4j}{(s + 2)^2 - (4j)^2} = \frac{2s + 4}{(s + 2)^2 + 16}$$

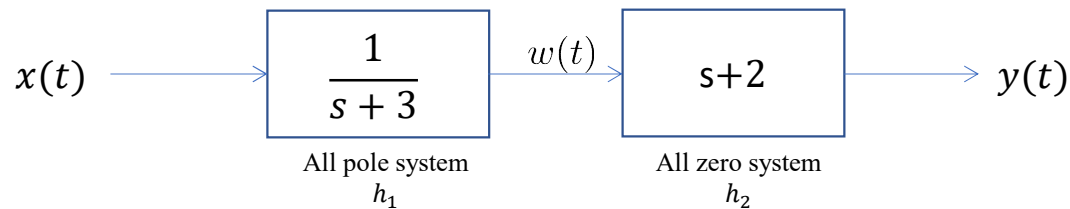
$$e^{-at} \cos(bt)u(t) \xrightarrow{L} \frac{s+a}{(s+a)^2+b^2}$$

$$H(s) = \frac{2(s + 2)}{(s + 2)^2 + 4^2} \xrightarrow{\text{Inverse Laplace}} 2e^{-2t} \cos(4t)u(t)$$

# Laplace Transform

Example with zero & pole:

$$H(s) = \frac{s + 2}{s + 3}$$

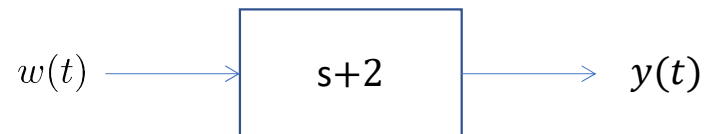
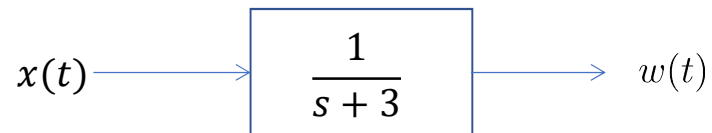


$$h_1(t) = e^{-t}u(t) : \quad \frac{X(s)}{s+3} = W(s)$$

$$x(t) = \frac{d}{dt}w(t) + 3w(t)$$

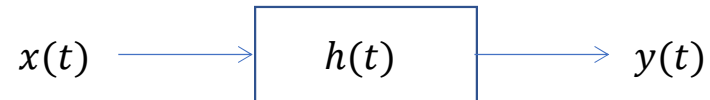
$$h_2(t) = \delta'(t) + 2\delta(t) : \quad Y(s) = (s+2)W(s)$$

$$y(t) = \frac{d}{dt}w(t) + 2w(t)$$



$$h(t) = h_1(t) * h_2(t) = e^{-t}u(t) * (2\delta(t) + \delta'(t)) = 2e^{-t}u(t) + \frac{d}{dt}(e^{-t}u(t))$$

# Laplace Transform



$$x(t) = e^{s_0 t}$$

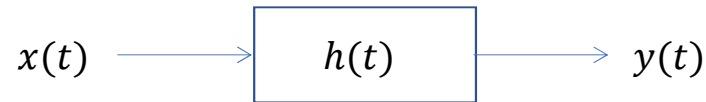
$$y(t) = H(s_0)e^{s_0 t}$$

Only for  $s_0$ s that are in ROC of  $H(s)$ , otherwise the output is infinity!

## Example:

$$x(t) = C \text{ (Constant)} \Rightarrow y(t) = C \times H(j0)$$

# Laplace Transform



$$x(t) = e^{s_0 t}$$

$$y(t) = H(s_0)e^{s_0 t}$$

Only for  $s_0$ s that are in ROC of  $H(s)$ , otherwise the output is infinity!

Note that  $e^{s_0 t}$ s are eigenfunctions of Laplace Transform!.

Difference between  $e^{s_0 t}$  and its causal part  $e^{s_0 t}u(t)$ ,  $s_0 = \alpha_0 + j\omega_0$ :

$$e^{s_0 t} \xrightarrow{\text{Laplace}} \delta(s - s_0)$$

$$ROC = \{s \mid Re\{s\} = Re\{s_0\} = \alpha_0\}$$

$$e^{s_0 t}u(t), \xrightarrow{\text{Laplace}} X_1(s) = \frac{1}{s - s_0}, ROC = \{s \mid Re\{s\} > \alpha_0\}$$

